# DELFT UNIVERSITY OF TECHNOLOGY FACULTY OF AEROSPACE ENGINEERING

Course Date	:	Stochastic Aerospace Systems (ae4-304) April 7, 2011 from 9:00 until 12:00 hr
Remarks	:	Write your name, initials and student number on your work Answer all questions in English or Dutch and mark all pages with your name.

The exam consists of 9 questions, a correctly answered question is rewarded with 10 points. The final mark is then:

$$1 \le 1 + 9 * (0 \dots 10) / 10 \le 10$$

# 1. CORRELATION FUNCTION (10 points)

Consider the output  $\bar{y}(t)$  of a first order system to a zero-mean, Gaussian white noise signal  $\bar{w}(t)$ . The first order system dynamics are given by the following transfer function:

$$H^y_w(s) = \frac{1}{1+\gamma s}$$

with time constant  $\gamma$ .

The auto-correlation of the output signal  $\bar{y}(t)$ ,  $K_{\bar{y}\bar{y}}(\tau)$  is illustrated in Figure 1, for three values of the time constant  $\gamma$  ( $\gamma_1$  (a),  $\gamma_2$  (b), and  $\gamma_3$  (c) ).

Which of the following statements is true? Explain your answer.

- $[\mathbf{a}] \quad \gamma_1 > \gamma_2 > \gamma_3$
- $[\mathbf{b}] \quad \gamma_1 < \gamma_2 < \gamma_3$
- [c] None of the auto-correlation functions represents the response of a first order system to a white noise input signal.

### 2. FOURIER TRANSFORM (10 points)

<u>Prove</u> Parceval's theorem:

$$\int_{-\infty}^{\infty} x^2(t) dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} |X(j\omega)|^2 d\omega$$

### 3. PROBABILITY DENSITY FUNCTION (10 points)

The random variable  $\bar{x}$  has a probability density function  $f_{\bar{x}}(x)$  as depicted in Figure 2.

What is the probability  $\Pr{\{\bar{x} \ge -1\}}$ ?

**[a]** 0.125



Figure 1: The auto-correlation  $K_{\bar{y}\bar{y}}(\tau)$  of the output signal  $\bar{y}(t)$  of a first order system driven by white noise.



Figure 2: The probability density function  $f_{\bar{x}}(x)$  of stochastic variable  $\bar{x}$ .

- **[b]** 0.275
- [c] 0.725
- [**d**] 0.750
- [e] 0.875
- **[f]** Not enough data available

### 4. FOURIER TRANSFORM (10 points)

Given a signal x(t). This signal is Fourier transformed, resulting in  $X(\omega)$ . The real part of  $X(\omega)$  is zero, the imaginary part of  $X(\omega)$  is non-zero.

Which of the following statements is true? Explain your answer.



Figure 3: Closed loop system configuration.

- [a] x(t) is odd.
- [b] x(t) is even.
- [c] x(t) is not odd, not even.
- [d] Without more information, one cannot determine whether any of the former answers are correct.

### 5. SYSTEM (10 points)

Given an closed loop system configuration shown in Figure 3. In this figure,  $H_1$ ,  $H_2$  and  $H_3$  are Linear Time-Invariant systems.

Signals  $\bar{n}_1$  and  $\bar{n}_2$  are zero-mean, uncorrelated white noise with intensities  $N_1$  and  $N_2$ , respectively. Signal  $\bar{u}$  is an arbitrary zero-mean input signal that is uncorrelated with  $\bar{n}_1$  and  $\bar{n}_2$ .

<u>Derive</u> an expression that relates the Power Spectral Density of the output signal  $\bar{y}$  to the PSDs of all input signals  $(\bar{u}, \bar{n}_1, \bar{n}_2)$  in the closed loop.

## 6. DISCRETE FOURIER TRANSFORM (10 points)

Consider two time-domain signals, sampled using the discretization time  $\Delta t$ ,  $x[n] = x[n \Delta t]$  and  $y[n] = y[n \Delta t]$ . The number of samples in both arrays is N.

[a] Given the (auto) circular covariance function,

$$C_{xx}[r] = \frac{1}{N} \sum_{n=0}^{N-1} x[n]x[n+r]$$

with r the separation sample, then <u>prove</u> that the (auto) Periodogram  $I_{xx}[k]$  equals

$$I_{xx}[k] = \frac{1}{N}X[-k]X[k] \tag{1}$$

by Discrete Fourier Transforming the (auto) circular covariance function  $C_{xx}[r]$  according to





$$I_{xx}[k] = \sum_{r=0}^{N-1} C_{xx}[r] e^{-j\frac{2\pi kr}{N}}$$
(2)

with in Eq. (1) X[k] the DFT of x[n], and in Eqs. (1) and (2) k the frequency counter.

- **[b]** <u>Prove</u>/explain that the auto Periodogram  $I_{xx}[k]$  is real-valued for all k.
- [c] Now consider the cross Periodogram  $I_{xy}[k]$  with

$$I_{xy}[k] = \frac{1}{N}X[-k]Y[k] \tag{3}$$

<u>Prove</u> that the cross Periodogram  $I_{xy}[k]$  is complex-valued for all k.

- [d] Briefly explain how the cross Periodogram  $I_{xy}[k]$  may be used in practical applications such as system-identification in the frequency-domain (that is, the estimation of the frequency response function  $H[k] \approx H(\omega)$  with  $\omega$  the circular frequency (in rad/s)).
- [e] What is the definition of the discrete frequency array  $\omega_k$ ? Use the frequency counter k, the number of samples N and the discretization time  $\Delta t$  for the definition.

## 7. SAMPLING (10 points)

A continuous-time (CT) signal y(t) is sampled, with a sampling frequency of 20 Hz, resulting in a discrete-time (DT) signal y[k].

The original CT Fourier transform of the signal,  $Y(\omega)$ , is illustrated in Figure 4.

Which of the following statements is true? Please explain your answer.

- [a] The CT Fourier transform of the discretized signal y[k] is shown in Figure 5(a).
- [b] The CT Fourier transform of the discretized signal y[k] is shown in Figure 5(b).
- [c] The CT Fourier transform of the discretized signal y[k] is shown in Figure 5(c).
- [d] None of the above is correct.



Figure 5: Three possible CT Fourier transforms of DT signal y[k].

# 8. LYAPUNOV (10 points)

Given a Linear Time-Invariant system with transfer function H(s):

$$H(s) = \frac{1}{\left(1 + 2\zeta \frac{s}{\omega_0} + \left(\frac{s}{\omega_0}\right)^2\right)}.$$

This system is driven by a signal that is zero for t < 0 and equals zero-mean white noise  $\bar{w}$  for t > 0 (the white noise intensity W equals 1). The system is at rest at t = 0.

<u>Prove</u> that the "steady state" variance of the output signal  $\bar{y}$ ,  $\sigma_{\bar{y}}^2$  equals  $\frac{\omega_0}{4\zeta}$ .

<u>Tip</u>: create the system state-space description (A, B, C, D), then solve the Lyapunov equation  $AC_{xx,ss} + C_{xx,ss}A^T + BWB^T = 0$ , with  $C_{xx,ss}$  the steady-state covariance of the state equation, and then compute  $C_{yy,ss}$ .

#### 9. SPECTRUM (10 points)

Given an LTI system, the transfer function of which equals:

$$H(s) = \frac{1}{1 + \frac{2\zeta}{\omega_0}s + \frac{1}{\omega_o^2}s^2},$$

with  $\omega_0 = 1$  rad/s and  $\zeta = 0.7$ .

This system is driven by a white noise input signal  $\bar{w}$  that has intensity W = 2,



Figure 6: Power spectral densities  $S_{\bar{y}\bar{y}}(\omega)$ .

resulting in an output signal  $\bar{y}$ .

Figure 6 shows four possible Power Spectral Densities  $S_{\bar{y}\bar{y}}(\omega)$ .

Which of the following statements is true? Explain your answer!

- [a] PSD (1) is the correct spectrum.
- [b] PSD (2) is the correct spectrum.
- [c] PSD (3) is the correct spectrum.
- [d] PSD (4) is the correct spectrum.
- [e] None of the spectra in Figure 6 is correct.